

Flexible Quantization

Bastiaan Kleijn KTH School of Electrical Engineering Stockholm



- Motivation for coding technologies
- Basic quantization and coding
- High-rate quantization theory



- Digital representation of signal
 - Sequence of samples with finite precision
 - Robust against distortion
 - Facilitates processing
- Basic rates for audio signals:
 - 48 kHz audio, 16 bits, stereo: 1536000 bits/second
 - 8 kHz speech, 16 bits: 128000 bits/second



High Rate is Expensive

- Transmission
 - Wired links
 - "Last mile"
 - Packet networks
 - Switching video
 - Wireless links
 - WiFi
 - Mobile telephony: coding was enabling technology
 - Secure communication
- Storage
 - Portable audio/video players
 - Output surveillance cameras



- Conventional circuit-switched networks
 - Virtually no bit errors, no loss
- Mobile networks
 - Reasonable cost and delay implies bit errors
- Packet networks
 - Reasonable cost and delay implies packet loss



Networks More Diverse

- How it was:
 - Single-paradigm network end-to-end
 - One service
- How it is:
 - Many paradigms in one composite network:
 - Circuit-switched network
 - Packet network
 - Wireless circuit-switched network
 - Wireless packet network
 - Many types of service
 - Range of quality-cost
 - Streaming, one-on-one communication



- New application (particular network, storage) appeared
- Study application requirements
- Design coder for application requirements
- Have competition between coder designs for conditions of application
- Select best coder
- No vision of integrated network



- Attributes of a coder
 - Rate
 - Quality (subjective), includes signal bandwidth
 - Delay
 - Robustness: bit errors and packet loss
 - Computational complexity
- Designs selected for one configuration of attributes
 - Associated with one network paradigm
 - Design effort irrelevant



- Implications of old-school design in new world:
 - Coders implicitly unable to adapt: codebooks
 - Transcoding
 - Performance unclear when applied to other conditions
 - GSM coder applied to packet networks
- New-school design
 - Goal: coders that can adapt in real-time to
 - Network conditions
 - Quality requirements
 - Near-optimal over large range of conditions
 - Employ high-rate quantization theory and more modeling



- Motivation for coding technologies
- Basic quantization and coding
- High-rate quantization theory



- Quantization: non-invertible mapping from Euclidian space \mathscr{R}^k to a countable set of points $\mathscr{C} = \{c_i\}$ that is a subset of \mathscr{R}^k
 - Quantization cell: $V_i = \{x \in \mathcal{R}^k : Q(x) = c_i\}$
 - "Inverse quantization" is misnomer



Example: Scalar Quantizer





Example: Vector Quantizer





- $V_i = \{x \in \mathscr{R}^k : Q(x) = c_i\}$ is a cell
 - Usually assumed convex: "regular" quantfzers
 - Cell = Voronoi region
- The quantization index i specifies the cell and the reconstruction point (often called the centroid)
- If the set of indices {i} is countable, the quantization index can generally be transmitted with a finite number of bits

$$x \quad \text{encoder} \quad i \quad \text{network} \quad i \quad \text{decoder} \quad c_i = \{c_i \in \mathcal{C}_i : x \in V_i\}$$



Example: Vector Quantizer





- Is it smart to simply transmit the index i? NO!
 (it is if index probability uniform)
- Apply lossless (entropy) coding to indices
 - Used to create ".zip"





- Code: the set of all codewords $\{w_i\}$
- Uniquely decodable code: can always reconstruct
- ~Minimum codeword length for uniquely decodable code: $l(w_i) = -\log_2(p_I(i))$ (follows from *Kraft inequality*)

• Entropy of the index:
$$H(I) = -\sum_{i} p_{I}(i) \log_{2}(p_{I}(i))$$

- Is ~minimum average rate needed for index
 - More accurately: $H(I) \le L < H(I) + 1$



Index resembles coin flips

$$H(I) = -\sum_{i} p_{I}(i) \log_{2}(p_{I}(i)) = -0.5 \log_{2}(0.5) - 0.5 \log_{2}(0.5) = 1 \text{ bits}$$

• Index resembles biased coin flips $H(I) = -0.25 \log_2(0.25) - 0.75 \log_2(0.75) = 0.811 \text{ bits}$ $= -0.05 \log_2(0.05) - 0.95 \log_2(0.95) = 0.286 \text{ bits}$



- To quantize we need to know with respect to what
- Optimal trade-off *distortion versus number of indices*
 - Constrained-resolution
 - Assumes codeword length is fixed
 - Generally short delay
 - Consistent with TDMA and FDMA, circuit-switched networks
 - The past
- Optimal trade-off distortion versus average rate
 - Constrained-entropy
 - Assumes only average codeword length matters
 - Often long delay
 - Consistent with CDMA and packet-switched networks
 - The future!



Old-School, Any-Rate Quantization

- Standard approach
 - Constrained-resolution
 - Stored codebooks
 - Codebooks trained with data
 - (Generalized) Lloyd algorithm (GLA), Bell Labs, 1958
 - / K-means algorithm



Is constrained-resolution





Lloyd Algorithm





Outcome Lloyd for Vector Quantizer







Practical (Discrete) Lloyd Algorithm

- Have database $\{x_m^k\}_{m \in M}$
- Encoder = partition = $\{V_j = \{x_i^k\}_{i \in \mathcal{J}(j)}: \bigcup V_j = \{x_i^k\}_{i \in I}\}$
- Decoder = codebook = $C_i = \{c_i\}$





• Optimize = minimize overall distortion:



Old-School, Any-Rate Quantization

- Is in your cell phone
- Constrained resolution (fixed number of cells/centroids)
- Works even at low rates
- Locally optimal
 - Distortion decreases each step
- Training computationally expensive: not in real time
 - Iterative solutions only
- Many variants:
 - Multi-stage
 - Tree
 - Constrained-entropy version (around 1990)



- Assume data density can be assumed constant within a cell (Bennett, 1948)
- Assume that notion *density of centroids* is meaningful
- Problem formulation
 - Given data density, distortion criterion, constraint
 - Find centroid density ("quantizer")
- Advantage of approach
 - Optimal quantizer can be computed analytically
 - Can be done in real-time



Distortion and Geometry: SQ

$$D_{i} = \frac{\int_{V_{i}} f_{X}(x)d(x,Q(x))dx}{\int_{V_{i}} f_{X}(x)dx} \approx \frac{f_{X}(x)\int_{V_{i}} (x-c_{i})^{2}dx}{f_{X}(x)\Delta_{i}}$$

$$= \frac{1}{\Delta_{i}} \int_{-\Delta_{i}/2}^{\Delta_{i}/2} x^{2}dx = \frac{\Delta_{i}^{2}}{12}$$

$$\Delta_{23}$$
• Scalar = cubic geometry



Mean distortion in cell i, r'th power criterion, per dim

$$D_{i} = \frac{\int_{V_{i}} f_{X}(x^{k})d(x^{k}, Q(x^{k}))dx^{k}}{\int_{V_{i}} f_{X}(x^{k})dx^{k}} \approx \frac{1}{kV_{i}} \int_{V_{i}} \|x^{k} - c_{i}^{k}\|_{r} dx^{k}$$
$$= V_{i}^{\frac{r}{k}} \frac{1}{k} V_{i}^{-\frac{r+k}{k}} \int_{V_{i}} \|x^{k} - c_{i}^{k}\|_{r} dx^{k} = V_{i}^{\frac{r}{k}} C(r, k, G(i))$$

• $C(r,k,G(i)) \approx C(r,k,G(x^k))$ is the inertial profile coefficient of quantization



• Scalar case:
$$C(r=2, k=1, G = \text{optimal}) \approx \frac{1}{12} = 0.0833$$

- cubic cells

• 2-D:
$$C(r=2, k=2, G = \text{optimal}) \approx \frac{5}{36\sqrt{5}} = 0.0802$$

– Hexagonal in sets of two dimensions

•
$$\infty$$
-D: $C(r = 2, k = \infty, G = \text{optimal}) \approx (2\pi e)^{-1} = 0.0585$
- Spherical cells

• VQ has space-filling advantage; 1.53 dB (= 0.25 bit)



CE Quantizers in 2D

• Two dimensions: square and hexagonal lattice







 What we have done: relate local geometry to local distortion

• Next step:

to relate distortion, rate and centroid density (and local geometry)

Centroid density: number of centroids/unit volume
 g(x^k)



- Apply lossless (entropy) coding to indices
 - Used to create ".zip"
- Rate is mean rate of codewords
 - Consistent with CDMA, packet networks, the future





• Constraint on index entropy

$$H(I) = -\sum_{i} p_{I}(i) \log(p_{I}(i))$$

$$= -\sum_{i} V_{i} f_{X}(c_{i}) \log(V_{i} f_{X}(c_{i}))$$

$$\approx -\int f_{X}(x^{k}) \left(\log(f_{X}(x^{k})) - \log(g_{X}(x^{k}))\right) dx^{k}$$

$$= h(X^{k}) + \int f_{X}(x^{k}) \log(g_{X}(x^{k})) dx^{k}$$

• Equivalent constraint

$$\int f_X(x^k) \log(g_X(x^k) dx^k) = \text{constant}$$



• Distortion:

$$D = \sum_{i} p_{I}(i)D_{i} = \sum_{i} p_{I}(i) V_{i}^{-\frac{r}{k}}C(r,k,G(i))$$

$$\approx C(r,k,G)\int f_{X}(x^{k})g^{\frac{r}{k}}(x^{k})dx$$

• Add Lagrange-multiplier term:

$$D \approx C(r,k,G) \int f_X(x^k) g^{-\frac{r}{k}}(x^k) dx$$

= $C(r,k,G) \int f_X(x^k) \left(g^{-\frac{r}{k}}(x^k) + \lambda \log(g(x^k)) \right) dx$
as get Euler-Lagrange equation: solve

• Minimize; get Euler-Lagrange equation; solve

$$f_X(x^k) \left(g^{-\frac{r+k}{k}}(x^k) + \lambda g^{-1}(x^k)) \right) = 0 \quad \rightarrow \qquad g(x^k) = \text{constant}!!$$



- For constrained-entropy quantization: simplest quantizer is best
 - All cells are same size and shape (not proven, that)
 - Facilitates low computational complexity quantizer
 - Can compute quantizer for given pdf and distortion
 - Does **not** mean entire encoder is low complexity!
- Somewhat non-intuitive:
 - Infinite number of cells/centroids!
 - Cell size independent of data density



CE Quantizers in 2D

• Two dimensions: square and hexagonal lattice







• Complete solution:

$$g(x^k) = \exp(H(I) - h(X^k))$$

- At a given distortion level the optimal centroid density:
 - increases with mean index rate
 - decreases with differential entropy of data (= complexity of data)
- Can adjust coder in real time!



• Relation distortion and rate (per dimension):

$$D = \sum_{i} p_{I}(i)D_{i} \approx C(r,k,G)\int f_{X}(x^{k})g(x^{k})^{-\frac{r}{k}}dx^{k}$$
$$= C(r,k,G)g(x^{k})^{-\frac{r}{k}}\int f_{X}(x^{k})dx^{k}$$
$$= C(r,k,G)\exp\left(-\frac{r}{k}\left(H(I)-h(X^{k})\right)\right)$$



• Divide distortions of SQ and VQ (Gray & Lookabough, 1989)

$$\frac{D_{SQ}}{D_{VQ}} = \frac{C(r, l, G) \exp\left(-r\left(H(I) - h(X^{1})\right)\right)}{C(r, k, G) \exp\left(-\frac{r}{k}\left(H(I) - h(X^{k})\right)\right)}$$
$$= \frac{C(r, l, G)}{C(r, k, G)} \exp\left(r\left(h(X^{1}) - \frac{1}{k}h(X^{k})\right)\right)$$

- Space-filling advantage
- Memory advantage (due to redundancy)

$$\rho = h(X^1) - \frac{1}{k}h(X^k)$$



- Uniform quantizer:
 - simple to implement
 - Small advantage from using best lattice
 - Somewhat more complicated

- Lossless coding is not easy:
 - Does not even exist in old-school quantization
 - Must know data density



- Lossless coding tries to reduce rate to index entropy
- Huffman code:
 - Table $i \rightarrow w_i$ based on probability distribution
 - Works on per-variable basis; high overhead
 - Simple to implement
- Arithmetic code:
 - Computes codewords for sequence of coefficients
 - Tricky to write program
 - Low overhead
 - Requires cumulative distribution function (cmf)
 - Often nontrivial to obtain cmf
 - Preferred method



- No significant commercial implementations as yet
- Quantizer and arithmetic coder are *computed; flexible*





- Difficult; simplify problem:
 - Density modeled as mixture $p_X(x^k) = \sum p_M(m) p_{X,m}(x^k)$
 - Interpretation: data fall in one of set of probabilities
 - Each mixture component is Gaussian (usually)
 - Know how to design quantizer for Gaussian
 - Symmetric
 - Just one design procedure needed for cmf computation
 - Encode which component you select then use corresponding quantizer



Gaussian mixture

• Four components:





- Not yet widely applied
 - Real-time adaptation not used
- Constrained entropy (constraint on average rate)



What Have We Learned

- Problem:
 - Have audio or video data (transformed or not)
 - Need to encode efficiently
- Old-School Solution
 - Good performance / not flexible
 - Constrained resolution
 - Codebook (often computationally expensive)
 - Commonly used
- New-World Approach
 - Good performance / can adapt in real-time
 - Constrained entropy; requires lossless coder (arithmetic coder)
 - Quantizer and arithmetic coder computed = flexible
 - Not yet ready



- Emphasis was on performance
- Emphasis is on flexibility (but no loss of performance)